

# Virtual Synchronous Machine Grid-Forming Inverter Model Specification (REGFM\_B1)

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# 1 Introduction

This report describes a generic virtual synchronous machine (VSM) grid-forming inverter (GFM) model—REGFM\_B1. The initial model specification was proposed by Pacific Northwest National Laboratory (PNNL), General Electric (GE), and Electric Power Research Institute (EPRI). Siemens Gamesa Renewable Energy (SGRE) also provided inputs to the specification. The model specification has been revised multiple times based on the discussions between all the contributors listed in this report. This work was funded by the Universal Interoperability for Grid-Forming Inverters (UNIFI) Consortium.

This generic model is developed to help the utility industry understand the concept of VSM GFMs. The model could be used to represent equipment for long-term planning studies where vendor-specific models are not available. As equipment mature and improve, generic models will be updated to capture the new functionalities of GFMs. It is not intended that these models will always remain representative of all future GFM technologies.

## 2 VSM GFM Model (REGFM\_B1)

The main circuit of a VSM GFM can be represented as a voltage source behind impedance as shown in Figure 1.

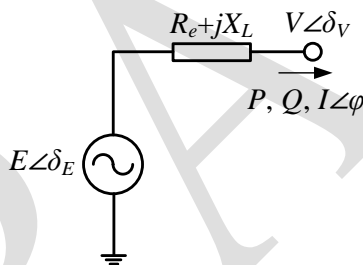


Figure 1 Voltage source representation of a grid-forming inverter

Figure 2 shows the virtual synchronous machine control block. In (1) and (2),  $P$  and  $I_d$  are per unit values on the system base, and  $P_{inv}$  and  $I_{dinv}$  are per unit values on the inverter rating base.

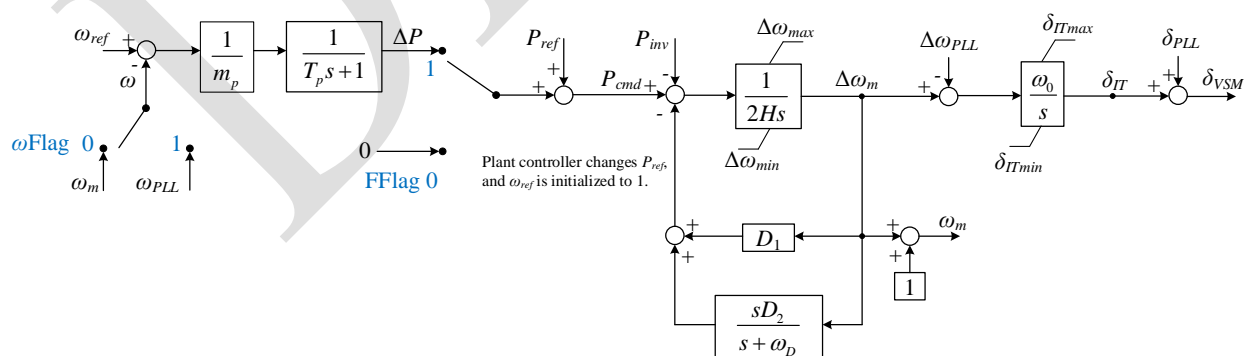
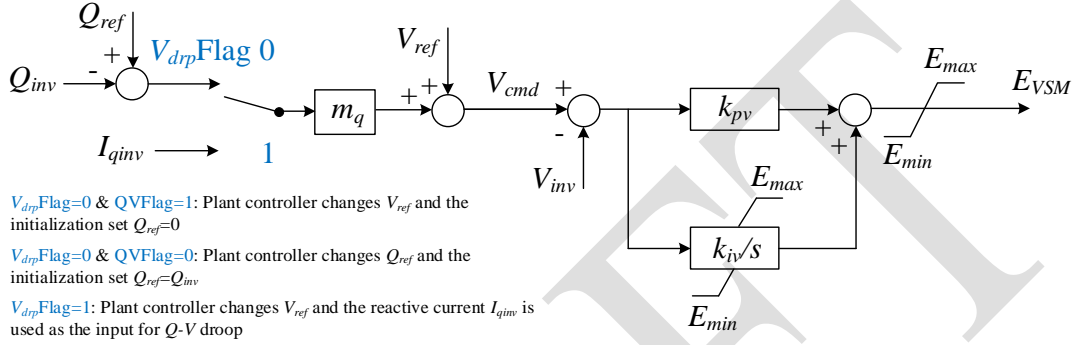


Figure 2 Virtual synchronous machine control

$$P_{inv} = \frac{1}{1+T_{pf}s} P \frac{S_{base}}{M_{base}} \quad (1)$$

$$I_{dinv} = \frac{1}{1+T_{lf}s} I_d \frac{S_{base}}{M_{base}} \quad (2)$$

Figure 3 shows the voltage control block. In (3) and (5),  $Q$  and  $I_q$  are per unit values on the system base, and  $Q_{inv}$  and  $I_{qinv}$  are per unit values on the inverter rating base.



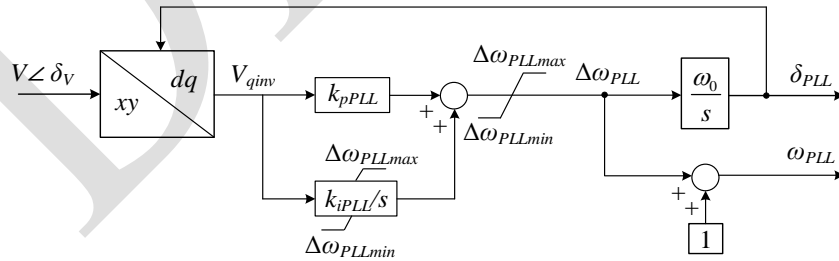
**Figure 3 Voltage control**

$$Q_{inv} = \frac{1}{1+T_{Qf}s} Q \frac{S_{base}}{M_{base}} \quad (3)$$

$$V_{inv} = \frac{1}{1+T_{Vf}s} V \quad (4)$$

$$I_{qinv} = \frac{1}{1+T_{lf}s} I_q \frac{S_{base}}{M_{base}} \quad (5)$$

Figure 4 shows the PLL control block, and the voltages and currents in the  $dq$  frame can be calculated using (6)–(9).



**Figure 4 PLL block**

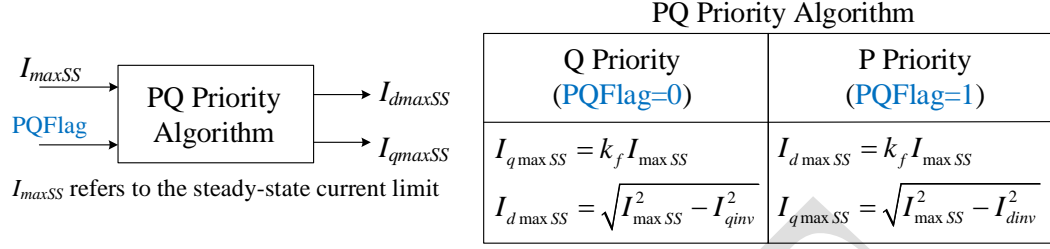
$$I_d = I_x \cos \delta_{PLL} + I_y \sin \delta_{PLL} \quad (6)$$

$$I_q = -I_x \sin \delta_{PLL} + I_y \cos \delta_{PLL} \quad (7)$$

$$V_d = V_x \cos \delta_{PLL} + V_y \sin \delta_{PLL} \quad (8)$$

$$V_q = -V_x \sin \delta_{PLL} + V_y \cos \delta_{PLL} \quad (9)$$

The PQ priority algorithm can be used to determine the steady-state active current limit  $I_{dmaxSS}$  and reactive current limit  $I_{qmaxSS}$ , as shown in Figure 5.



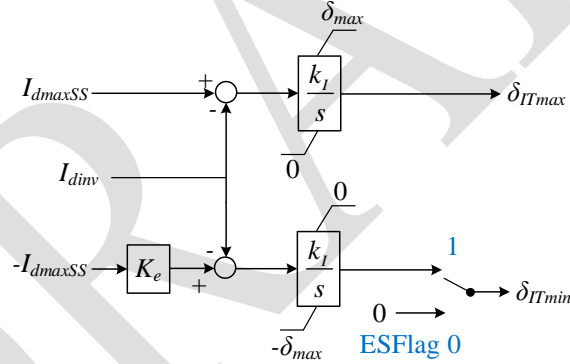
**Figure 5 PQ priority algorithm to determine the steady-state maximum active and reactive currents**

The steady-state reactive current  $I_{qinv}$  can be limited by reducing the internal voltage magnitude  $E$  using the algorithm described in (10) and (11).

$$E_{\min} = \sqrt{(V_{inv} - I_{q\ max\ SS} X_L)^2 + (I_{dinv} X_L)^2} \quad (10)$$

$$E_{\max} = \sqrt{(V_{inv} + I_{q\ max\ SS} X_L)^2 + (I_{dinv} X_L)^2} \quad (11)$$

The steady-state active current  $I_{dinv}$  can be limited by using the control block described in Figure 6. (12) describes how to calculate  $\delta_{max}$ .

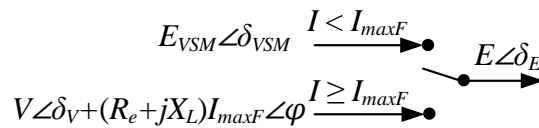


**Figure 6 Active current limiting block**

$$\delta_{\max} = \sin^{-1}(X_L I_{\ max\ SS}) \quad (12)$$

Equation (13) and Figure 7 describe the transient current limiting function. When the transient current limiting is activated, the output current is limited algebraically by the network solution. Note that  $I_{maxF}$  should be set larger than  $I_{maxSS}$ .

$$I \angle \varphi = \frac{E_{VSM} \angle \delta_{VSM} - V \angle \delta_V}{R_e + jX_L} \quad (13)$$



**Figure 7 Transient current limiting function**

**Table 1 Model Main Circuit and Controller Parameters**

Symbol	Description	Unit	Example Value
$\omega$ Flag	A flag to select whether to use the measured frequency from PLL as the input for the P-f droop.	NA	0
$V_{drp}$ Flag	A flag to select whether to use $Q$ or $I_q$ as the input of the $Q$ - $V$ droop control.	NA	0
QVFlag	A flag to determine whether $Q_{ref}$ or $V_{ref}$ should be used to interact with the plant controller.	NA	1
PQFlag	A flag to determine whether P priority (PQFlag=1) or Q priority (PQFlag=0) should be selected.	NA	0
FFlag	A flag to determine whether the power-frequency droop is enabled (FFlag=1) or disabled (FFlag=0).	NA	1
ESFlag	A flag to determine if the model represents a battery source (ESFlag=1) or a non-battery source (ESFlag=0).	NA	1
$R_e$	Inverter coupling resistance. ( $0 \text{ pu} \leq R_e \leq \frac{1}{4}X_L$ )	pu	0
$X_L$	Inverter coupling reactance. ( $0.04 \text{ pu} \leq X_L \leq 0.4 \text{ pu}$ )	pu	0.1
$m_q$	$Q$ - $V$ droop gain. When $V_{drpflag}=1$ , $m_q$ represents a per unit virtual impedance.	pu	0.05
$k_{pv}$	Proportional gain of the voltage controller	pu	0
$k_{iv}$	Integral gain of the voltage controller	pu/s	5
$m_p$	$P$ - $f$ droop gain	pu	0.02
$\Delta\omega_{max}$	Upper limit of $\Delta\omega_m$	pu	0.05
$\Delta\omega_{min}$	Lower limit of $\Delta\omega_m$	pu	-0.05
$k_{pPLL}$	Proportional gain of PLL	pu	0.265
$k_{iPLL}$	Integral gain of PLL	pu/s	2.65
$\Delta\omega_{PLLmax}$	Upper limit of the PLL output	pu	0.2
$\Delta\omega_{PLLmin}$	Lower limit of the PLL output	pu	-0.2
$T_p$	Time constant of the low-pass filter in the VSM control block. ( $T_p \geq 0$ )	s	0
$H$	Inertia time constant	s	0.5
$D_1$	Damping	pu	0
$D_2$	Transient damping	pu	100
$\omega_D$	Angular frequency of the washout block	pu	50
$I_{maxSS}$	Steady-state current limit. ( $I_{maxSS} \leq 0$ is treated as $1/X_L$ ) ( $I_{maxSS} \leq 1/X_L$ ) ( $I_{maxSS} \leq I_{maxF}$ )	pu	1
$k_f$	A factor to determine $I_{qmax}$ (PQFlag=0) or $I_{dmax}$ (PQFlag=1). If $k_f=0$ , the software should reset it to be 1 and generate a warning message.	NA	0.9
$k_I$	Integral gain for the active current limiting loop	pu/s	2
$I_{maxF}$	Transient current limit	pu	1.5
$T_{pf}$	Time constant of the low-pass filter for active power measurement	s	0.02 s
$T_{Qf}$	Time constant of the low-pass filter for reactive power measurement	s	0.02 s
$T_{Vf}$	Time constant of the low-pass filter for voltage measurement	s	0.02 s
$T_{If}$	Time constant of the low-pass filter for current measurement	s	0.02 s
$K_e$	Scalar on $I_{dmax}$ for negative active steady-state current limitation ( $0 \leq K_e \leq 1.0$ )	NA	1
$\omega_0$	Rated angular frequency. Software tools typically have this value specified in the solution environment, so this will not be listed as an input parameter of the model	rad/s	376.99

## References

Larsen EV, Delmerico RW, inventors; General Electric Co, assignee. Battery energy storage power conditioning system. United States patent US 5,798,633. 1998 Aug 25.

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